

1. 3.4-3 $H(f) = \frac{1}{1+j2\pi f}$, $h(t) = e^{-t}u(t)$

(i) $x(t) = 1 + 2\delta(t-t_0) - \cos(\omega_0 t) + \sum_{i=1}^n A_i e^{-a_i(t-t_i)} u(t-t_i)$, $a_i \neq 1$

$$\Rightarrow X(f) = \delta(f) + 2e^{-j2\pi f t_0} - \frac{1}{2} \delta(f - \frac{\omega_0}{2\pi}) - \frac{1}{2} \delta(f + \frac{\omega_0}{2\pi}) + \sum_{i=1}^n A_i \frac{1}{a_i + j2\pi f} e^{-j2\pi f t_i}$$

$$\Rightarrow Y(f) = H(f)X(f) = \delta(f) \cdot \frac{1}{1+j2\pi f} + \frac{2e^{-j2\pi f t_0}}{1+j2\pi f} - \frac{1}{2} \frac{1}{1+j\omega_0} \delta(f - \frac{\omega_0}{2\pi}) - \frac{1}{2} \frac{1}{1-j\omega_0} \delta(f + \frac{\omega_0}{2\pi}) + \sum_{i=1}^n \frac{A_i e^{-j2\pi f t_i}}{(a_i + j2\pi f)(1+j2\pi f)}$$

$$= \delta(f) + \frac{2e^{-j2\pi f t_0}}{1+j2\pi f} - \frac{1}{2} \frac{1}{1+j\omega_0} \delta(f - \frac{\omega_0}{2\pi}) - \frac{1}{2} \frac{1}{1-j\omega_0} \delta(f + \frac{\omega_0}{2\pi}) + \sum_{i=1}^n \left[\frac{A_i}{a_i + j2\pi f} + \frac{(-A_i)}{1+j2\pi f} \right] e^{-j2\pi f t_i}$$

$$\Rightarrow y(t) = 1 + 2e^{-(t-t_0)} u(t-t_0) - \frac{1}{2} \frac{e^{j\omega_0 t}}{1+j\omega_0} - \frac{1}{2} \frac{e^{-j\omega_0 t}}{1-j\omega_0} + \sum_{i=1}^n \left[\left(\frac{A_i}{1-a_i} \right) e^{-a_i(t-t_i)} - \left(\frac{A_i}{1-a_i} \right) e^{-(t-t_i)} \right] u(t-t_i)$$

$$= 1 + 2e^{-(t-t_0)} u(t-t_0) - \frac{\cos(\omega_0 t) + \omega_0 \sin(\omega_0 t)}{1+\omega_0^2} + \sum_{i=1}^n \left(\frac{A_i}{1-a_i} \right) \left(e^{-a_i(t-t_i)} - e^{-(t-t_i)} \right) u(t-t_i)$$

(ii) direct calculation :

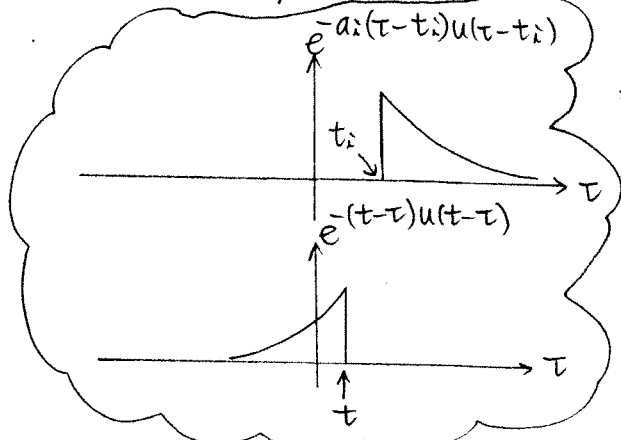
① $e^{-t}u(t) * 1 = \int_{-\infty}^{\infty} e^{-\tau} u(\tau) 1 d\tau = \int_0^{\infty} e^{-\tau} d\tau = -[e^{-\infty} - e^{-0}] = 1$

② $e^{-t}u(t) * 2\delta(t-t_0) = \int_0^{\infty} e^{-\tau} 2\delta(t-\tau-t_0) d\tau = 2e^{-(t-t_0)} u(t-t_0)$

③ $e^{-t}u(t) * \cos(\omega_0 t) = \frac{1}{2} \int_0^{\infty} e^{-\tau} e^{j\omega_0(t-\tau)} d\tau + \frac{1}{2} \int_0^{\infty} e^{-\tau} e^{-j\omega_0(t-\tau)} d\tau$

$$= \frac{1}{2} \frac{e^{j\omega_0 t}}{1+j\omega_0} + \frac{1}{2} \frac{e^{-j\omega_0 t}}{1-j\omega_0} = \frac{\cos(\omega_0 t) + \omega_0 \sin(\omega_0 t)}{1+\omega_0^2}$$

④ $e^{-t}u(t) * A_i e^{-a_i(t-t_i)} u(t-t_i) = A_i \int_{-\infty}^{\infty} e^{-a_i(\tau-t_i)} u(\tau-t_i) e^{-(t-\tau)} u(t-\tau) d\tau$



$$= \begin{cases} A_i \int_{t_i}^t e^{-a_i(\tau-t_i)} e^{-(t-\tau)} d\tau, & t > t_i \\ 0, & t \leq t_i \end{cases}$$

$$= \begin{cases} \frac{A_i}{1-a_i} e^{-a_i(t-t_i)} - \frac{A_i}{1-a_i} e^{-(t-t_i)}, & t > t_i \\ 0, & t \leq t_i \end{cases}$$

$$= \left(\frac{A_i}{1-a_i} \right) \left(e^{-a_i(t-t_i)} - e^{-(t-t_i)} \right) u(t-t_i)$$

①②③④ $\Rightarrow y(t) = 1 + 2e^{-(t-t_0)} u(t-t_0) - \frac{\cos(\omega_0 t) + \omega_0 \sin(\omega_0 t)}{1+\omega_0^2} + \sum_{i=1}^n \left(\frac{A_i}{1-a_i} \right) \left(e^{-a_i(t-t_i)} - e^{-(t-t_i)} \right) u(t-t_i)$

2. 3.5-1

The impulse response $h(t)$ of a causal system has to be zero $\forall t < 0$.

(a) Yes, $\because e^{-at} u(t) = 0, \forall t < 0$.

(b) No. There exists $t_1 < 0$ such that $e^{-a|t_1|} \neq 0$

(c) Yes, $\because e^{-a(t-t_0)} u(t-t_0) = 0, \forall t < 0 \quad (t_0 \geq 0)$

(d) No. There exists $t_1 < 0$ such that $\text{sinc}(at_1) \neq 0$.

(e) No. There exists $t_1 < 0$ such that $\text{sinc}(a(t_1-t_0)) \neq 0$.

3. 3.5-2 $H(f) = e^{-(2\pi k|f| + j2\pi f t_0)}$

$$\begin{aligned} \textcircled{1} h(t) &= \int_{-\infty}^{\infty} H(f) e^{j2\pi f t} df = \int_0^{\infty} e^{-2\pi k f} \cdot e^{j2\pi f(t-t_0)} df + \int_{-\infty}^0 e^{2\pi k f} \cdot e^{j2\pi f(t-t_0)} df \\ &= \frac{1}{-2\pi k + j2\pi(t-t_0)} (-1) + \frac{1}{2\pi k + j2\pi(t-t_0)} = \frac{4\pi k}{4\pi^2 k^2 + 4\pi^2(t-t_0)^2} \\ &= \frac{k}{\pi k^2 + \pi^2(t-t_0)^2} \end{aligned}$$

$\Rightarrow h(t)$ can be zero for some $t < 0 \Rightarrow$ noncausal
and hence physically unrealizable.
(Time-domain criterion)

A reasonable criterion: $h(0) = 1\%$ of peak value which is $\frac{k}{\pi k^2} = \frac{1}{\pi k}$

$$\text{i.e., } \frac{k}{\pi k^2 + \pi^2 t_0^2} = 0.01 \cdot \frac{1}{\pi k} \Rightarrow t_0 = k \sqrt{\frac{99}{\pi}} \approx 5.6136 k$$

This filter can be made approximately realizable by choosing a sufficiently large t_0 .

② Frequency domain (Paley-Wiener) criterion:

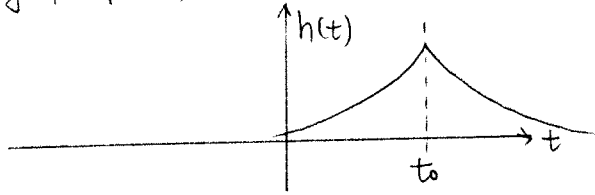
$$\int_{-\infty}^{\infty} \frac{|\ln|H(f)||}{4\pi^2 f^2 + 1} df = \int_{-\infty}^{\infty} \frac{|\ln e^{-(2\pi k|f|)}|}{4\pi^2 f^2 + 1} df = \int_{-\infty}^{\infty} \frac{|2\pi k f|}{4\pi^2 f^2 + 1} df = 2 \int_0^{\infty} \frac{2\pi k f}{4\pi^2 f^2 + 1} df$$

$$\uparrow = 2k \int_0^{\frac{\pi}{2}} \frac{\frac{\tan\theta}{2\pi} \sec^2\theta d\theta}{\tan^2\theta + 1} = \frac{k}{\pi} \int_0^{\frac{\pi}{2}} \tan\theta d\theta = \frac{k}{\pi} \left(-\ln|\cos\theta| \right) \Big|_0^{\frac{\pi}{2}} = \infty \Rightarrow \text{physically unrealizable}^*$$

$$\text{let } 2\pi f = \tan\theta, 2\pi df = \sec^2\theta d\theta$$

4. 3.5-3 $H(f) = \frac{2\beta}{(2\pi f)^2 + \beta^2} e^{-j2\pi f t_0}$

By Table 3.1 and time-shifting property, $h(t) = e^{-\beta|t-t_0|}$, which is noncausal and hence unrealizable.



The exponential decays to 1.8% at 4 times constants. Hence $t_0 = 4\beta^{-1}$ is a reasonable choice to make this filter approximately realizable.

5. 3.6-1 (a) $H(f) = e^{-j2\pi f t_0 - jk \sin 2\pi f T}$, $k \ll 1$
 $\approx e^{-j2\pi f t_0} [1 - jk \sin(2\pi f T)] = e^{-j2\pi f t_0} - \frac{k}{2} e^{-j2\pi f(t_0-T)} + \frac{k}{2} e^{-j2\pi f(t_0+T)}$

$$\Rightarrow h(t) \approx \delta(t-t_0) + \frac{k}{2} \delta(t-t_0-T) - \frac{k}{2} \delta(t-t_0+T)$$

$$\Rightarrow \text{Under this approximation, } y(t) = q(t-t_0) + \frac{k}{2} [q(t-t_0-T) - q(t-t_0+T)]$$

(b) Under the approximation in (a), $h(t)$ is a linear combination of linear distortion impulse responses. According to 3.6 of textbook, The distortion resulting from $h(t)$ is undesirable in a TDM system since the pulse spreading causes interference with a neighboring pulse (i.e. crosstalk).

For an FDM system, $h(t)$ causes distortion in each multiplexed signal while no interference occurs with a neighboring channel, because in FDM each multiplexed signal occupies a band which is not occupied by any other FDM signals.

$$6. \quad h_T(t) = \text{rect}\left(\frac{t-T/2}{T}\right) - \text{rect}\left(\frac{t+T/2}{T}\right)$$

$$\Rightarrow H_T(f) = T \text{sinc}(\pi f T) e^{-j2\pi f T/2} - T \text{sinc}(\pi f T) e^{j2\pi f T/2}$$

$$= T \text{sinc}(\pi f T) (-2j) \sin(\pi f T)$$

$$= T \frac{\sin(\pi f T)}{\pi f T} (-2j) \sin(\pi f T)$$

$$= \frac{2}{j\pi f} \sin^2(\pi f T)$$

$$= \frac{2}{j\pi f} \cdot \frac{1}{2} [1 - \cos(2\pi f T)] = \frac{1}{j\pi f} - \frac{\cos(2\pi f T)}{j\pi f} \neq \frac{1}{j\pi f}$$

We observe $H_T(f)$ cannot be determined as $T \rightarrow \infty$.

We can express $h_T(t)$ in another way :

$$h_T(t) = \text{rect}\left(\frac{t-T/2}{T}\right) - \text{rect}\left(\frac{t+T/2}{T}\right)$$

$$= \frac{1}{2} (\text{sgn}(t) - \text{sgn}(t-T)) - \frac{1}{2} (\text{sgn}(t+T) - \text{sgn}(t))$$

$$= \text{sgn}(t) - \underbrace{\frac{1}{2} (\text{sgn}(t-T) + \text{sgn}(t+T))}_{\neq 0 \text{ for all positive } T}$$

$$\neq \text{sgn}(t), \forall T > 0$$

$$\text{Hence } H_T(f) \neq \frac{1}{j\pi f}, \forall T > 0$$

7. 3.6-3 $y(t) = x(t) + 0.001x^2(t)$

(a) $x(t) = \frac{1000}{\pi} \text{sinc}(1000t) \Rightarrow y(t) = \frac{1000}{\pi} \text{sinc}(1000t) + 0.001 \left(\frac{1000}{\pi} \right)^2 \text{sinc}^2(1000t)$
 $= \frac{1000}{\pi} \text{sinc}(1000t) + \frac{1000}{\pi^2} \text{sinc}^2(1000t)$

By Fourier Transform Pairs : $\text{sinc}(\pi t) \Leftrightarrow \text{rect}(f)$
 $\text{sinc}^2(\pi t) \Leftrightarrow \Delta\left(\frac{f}{2}\right)$

We have : $Y(f) = \frac{1000}{\pi} \cdot \frac{\pi}{1000} \text{rect}\left(\frac{f}{1000}\right) + 0.001 \cdot \left(\frac{1000}{\pi}\right)^2 \cdot \frac{\pi}{1000} \Delta\left(\frac{f}{2 \cdot \frac{1000}{\pi}}\right)$
 $= \text{rect}\left(\frac{f}{1000}\right) + \frac{1}{\pi} \Delta\left(\frac{f}{2 \cdot \frac{1000}{\pi}}\right)$

(b) From (a), the bandwidth of $x(t)$ is $\frac{500}{\pi}$ and the bandwidth of $y(t)$ is $\frac{1000}{\pi}$. Thus that the bandwidth of the output signal is twice that of the input signal is verified.

(c) For general $x(t)$, it cannot be recovered without distortion since the signal squaring introduces nonlinear distortion that occupies the bandwidth which is overlapped with the bandwidth of original $x(t)$.

8. 3.7-1 We know $\text{sinc}^2(\pi t) \Leftrightarrow \Delta\left(\frac{f}{2}\right)$ Hence, $\text{sinc}^2(kt) = \text{sinc}^2\left(\pi t \frac{k}{\pi}\right) \Leftrightarrow \frac{\pi}{k} \Delta\left(\frac{f}{2 \cdot \frac{k}{\pi}}\right)$

$\int_{-\infty}^{\infty} \text{sinc}^4(kt) dt = \int_{-\infty}^{\infty} \left| \text{sinc}^2(kt) \right|^2 dt \underset{\text{Parseval's Theorem}}{=} \int_{-\infty}^{\infty} \left| \frac{\pi}{k} \Delta\left(\frac{f}{2 \cdot \frac{k}{\pi}}\right) \right|^2 df = \frac{\pi^2}{k^2} \int_{-\frac{k}{\pi}}^{\frac{k}{\pi}} \left| 1 - |f| \frac{\pi}{k} \right|^2 df$

$\rightarrow = \frac{2\pi^2}{k^2} \int_0^{\frac{k}{\pi}} \left(1 - f \frac{\pi}{k} \right)^2 df = \frac{2\pi^2}{k^2} \int_0^{\frac{k}{\pi}} \left(1 - 2f \frac{\pi}{k} + f^2 \frac{\pi^2}{k^2} \right) df = \frac{2\pi^2}{k^2} \left(f - \frac{\pi}{k} f^2 + \frac{1}{3} \frac{\pi^2}{k^2} f^3 \right) \Big|_0^{\frac{k}{\pi}}$
 $= \frac{2\pi^2}{k^2} \left(\frac{k}{\pi} - \frac{\pi}{k} \frac{k^2}{\pi^2} + \frac{1}{3} \frac{\pi^2}{k^2} \frac{k^3}{\pi^3} \right)$
 $= \frac{2\pi}{3k}$

9. 3.7-4

$$g(t) = \frac{2a}{t^2 + a^2}$$

From table 3.1, we know $x(t) = e^{-a|t|}$, $a > 0 \Rightarrow X(f) = \frac{2a}{a^2 + (2\pi f)^2}$

By duality property, $X(t) \Leftrightarrow x(-f)$, i.e., $\frac{2a}{a^2 + (2\pi t)^2} \Leftrightarrow e^{-a|-f|} = e^{-a|f|}$, $a > 0$

$$\Rightarrow \frac{2a}{t^2 + a^2} = \frac{2a}{\left(\frac{t}{2\pi} \cdot 2\pi\right)^2 + a^2} \Leftrightarrow 2\pi e^{-a|2\pi f|}, a > 0$$

$$\begin{aligned} \Rightarrow E_g &= \int_{-\infty}^{\infty} |G(f)|^2 df = 4\pi^2 \int_{-\infty}^{\infty} e^{-2a|2\pi f|} df = 4\pi^2 \int_{-\infty}^{\infty} e^{-4\pi a|f|} df = 8\pi^2 \int_0^{\infty} e^{-4\pi a f} df \\ &= 8\pi^2 \cdot \frac{1}{4\pi a} = \frac{2\pi}{a} \end{aligned}$$

$$E_{g,B} = \int_{-B}^B |G(f)|^2 df = 8\pi^2 \int_0^B e^{-4\pi a f} df = \frac{2\pi}{a} [1 - e^{-4\pi a B}] = \frac{2\pi}{a} \times 0.99$$

$$\Rightarrow e^{-4\pi a B} = 0.01 \Rightarrow B = \frac{1}{4\pi a} \ln 100 \simeq \frac{0.3665}{a} \text{ Hz}$$

10. 3.7-5

Let $g^2(t) \Leftrightarrow A(f)$, $y(t) \Leftrightarrow Y(f)$. We have $Y(f) = A(f) \text{rect}\left(\frac{f}{2\Delta f}\right)$

Because $\text{sinc}(\pi t) \Leftrightarrow \text{rect}(f)$

$$\Rightarrow \text{sinc}(\pi t 2\Delta f) \Leftrightarrow \frac{1}{2\Delta f} \text{rect}\left(\frac{f}{2\Delta f}\right)$$

$$\Rightarrow 2\Delta f \text{sinc}(\pi t 2\Delta f) \Leftrightarrow \text{rect}\left(\frac{f}{2\Delta f}\right)$$

When Δf is very small, $Y(f) \approx A(0) \text{rect}\left(\frac{f}{2\Delta f}\right)$

$$\text{Hence } y(t) \approx A(0) 2\Delta f \text{sinc}(\pi t 2\Delta f)$$

$$= A(0) 2\Delta f \frac{\sin(\pi t 2\Delta f)}{\pi t 2\Delta f} \approx A(0) 2\Delta f \frac{\pi t 2\Delta f}{\pi t 2\Delta f} = A(0) 2\Delta f$$

$\sin(x) \approx x$
when x is very small

On the other hand,

$$A(f) = \int_{-\infty}^{\infty} g^2(t) e^{-j2\pi f t} dt \Rightarrow A(0) = \int_{-\infty}^{\infty} g^2(t) dt = E_g$$

$$\text{Therefore } y(t) \approx A(0) 2\Delta f = 2E_g \Delta f$$