

## ENEE 630 Homework 2<sup>1</sup>

### Material Covered: Polyphase Representation, Uniform DFT Filter Banks, Analysis/Synthesis system

**Problem 1** Consider the structure shown in Fig P-1(a), where  $W$  is the  $3 \times 3$  DFT matrix.

This is a three channel synthesis bank with three filters  $F_0(z)$ ,  $F_1(z)$ ,  $F_2(z)$ . (For example  $F_0(z) = Y(z)/Y_0(z)$  with  $y_1(n)$  and  $y_2(n)$  set to zero.)

- a) Let  $R_0(z) = 1 + z^{-1}$ ,  $R_1(z) = 1 - z^{-2}$ ,  $R_2(z) = 2 + 3z^{-1}$ . Find expressions for the three synthesis filters  $F_0(z)$ ,  $F_1(z)$ ,  $F_2(z)$ .
- b) The magnitude response of  $F_1(z)$  is sketched in Fig. P-1(b). Plot the responses  $|F_0(e^{j\omega})|$  and  $|F_2(e^{j\omega})|$ . Does the relation between  $F_0(z)$ ,  $F_1(z)$ , and  $F_2(z)$  depend on choices of  $R_k(z)$ ?

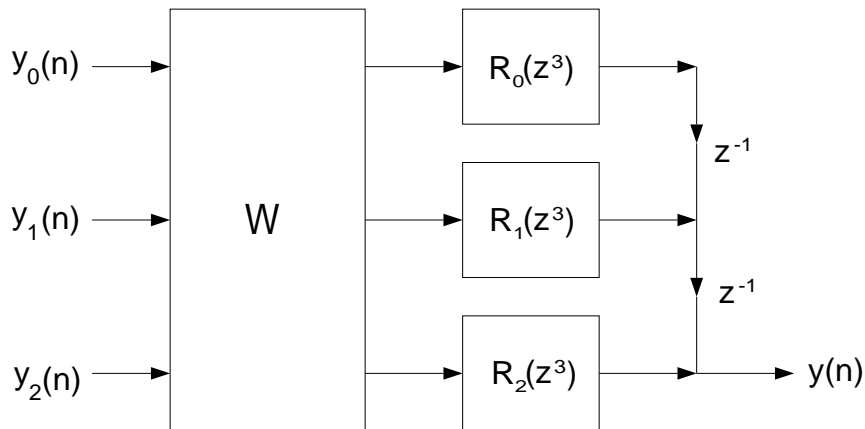


Figure: P-1(a)

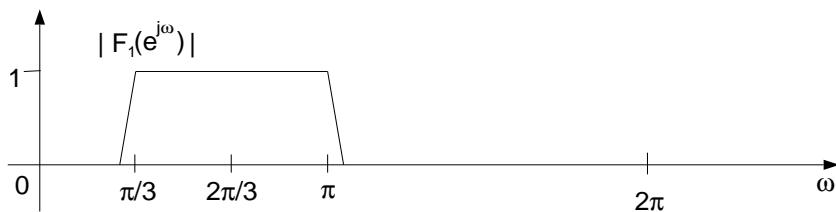


Figure: P-1(b)

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**Problem 2** For a uniform DFT analysis bank, we know that the filters are related by  $H_k(z) = H_0(zW^k)$ ,  $0 \leq k \leq M - 1$ , with  $W = e^{-j2\pi/M}$ . Let  $M = 5$  and define two new transfer functions  $G_1(z) = H_1(z) + H_4(z)$  and  $G_2(z) = H_2(z) + H_3(z)$ . Let  $h_0(n)$  denote the impulse response of  $H_0(z)$ , assumed to be real for all  $n$ .

- Are  $h_k(n)$ ,  $1 \leq k \leq 4$  real for all  $n$ ?
- Express the impulse responses  $g_1(n)$  and  $g_2(n)$  of  $G_1(z)$  and  $G_2(z)$  in terms of  $h_0(n)$ . Are  $g_1(n)$  and  $g_2(n)$  real for all  $n$ ?
- Let  $|H_0(e^{j\omega})|$  be as shown in Fig. P-2. Plot the responses  $|G_1(e^{j\omega})|$  and  $|G_2(e^{j\omega})|$ , for  $0 \leq \omega \leq 2\pi$ . Does  $|G_2(e^{j\omega})|$  necessarily look ‘good’ in its passband?



Figure : P-2

**Problem 3** Let  $H(z) = \sum_{n=0}^N h(n)z^{-n}$  with  $h(n) = h(N - n)$ . Consider the Type-1 polyphase representation for  $H(z)$  with  $M$  polyphase components. The symmetry of  $h[n]$  reflects into the coefficients of the polyphase components  $E_l(z)$  as follows: there exists an integer  $m_0$  (with  $0 \leq m_0 \leq M - 1$ ) such that  $e_k[n]$  is the image of  $e_{m_0-k}[n]$  for  $0 \leq k \leq m_0$ , and  $e_k[n]$  is the image of  $e_{M+m_0-k}[n]$  for  $m_0 + 1 \leq k \leq M - 1$ .

- Take an example of a 7<sup>th</sup>-order  $H(z)$  and verify the above statement for  $M = 3$ . What is  $m_0$ ? How about when  $M = 4$ ?
- Prove the above statement. Find out how  $m_0$  is related to  $N$  and  $M$ .

**Problem 4** Consider the analysis/synthesis system in Fig. P-4.

- a) Let the analysis filters be  $H_0(z) = 1 + 3z^{-1} + 0.5z^{-2} + z^{-3}$  and  $H_1(z) = H_0(-z)$ . Find causal stable IIR filters  $F_0(z)$  and  $F_1(z)$  such that  $\hat{x}(n)$  agrees with  $x(n)$  except for a possible delay and (nonzero) scale factor.
- b) Let  $H_0(z) = 1 + 2z^{-1} + 3z^{-2} + 2z^{-3} + z^{-4}$ , and  $H_1(z) = H_0(-z)$ . Find causal FIR filters  $F_0(z)$  and  $F_1(z)$  such that  $\hat{x}(n)$  agrees with  $x(n)$  except for a possible delay and (nonzero) scale factor.

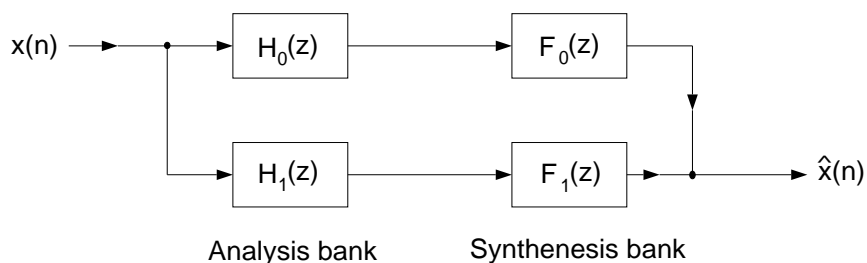


Figure : P-4

(Hint: Polyphase decomposition may help to solve this problem. Review of complementary filters might help as well. And there is a counter part of Euclid's theorem in polynomial form for two relatively prime polynomials.)